

A Class of Nonlinear H_∞ Controller Design via State Feedback for Disturbance Attenuation *

LI Xi, TANG Xiaoqi, ZHOU Yunfei and CHEN Jihong

(National Engineering & Technology Research Center of Numerical Control · Wuhan, 430074, P. R. China)

Abstract: A class of disturbance attenuation approach is proposed for affine nonlinear systems, which yields the local stable systems of these problems via state feedback standard H_∞ control for unknown finite bounded periodical disturbance. The method for these problems is showed by using dissipation inequality, and the parameters of controller are derived.

Key words: nonlinear system; H_∞ -controller; state feedback; disturbance attenuation

Document code: A

一类抑制扰动的 H_∞ 状态反馈非线性控制器设计

李 曦 唐小琦 周云飞 陈吉红

(国家数控系统工程技术研究中心·武汉, 430074)

摘要: 研究了一类状态可测仿射非线性系统的扰动抑制问题, 针对局部渐近稳定的非线性系统受到未知有界的周期性扰动, 以 H_∞ 状态反馈标准控制方法为基础, 给出这类问题的设计方法, 得到了控制器的参数化描述。

关键词: 非线性系统; H_∞ 控制器; 状态反馈; 干扰抑制

1 Introduction

A popular method to solve the standard H_∞ control problem is the state feedback approached and developed in [1]. Recently, attention was extended to H_∞ robust performance problem for nonlinear systems and there has been considerable progress in the design of robust H_∞ controller for nonlinear systems. The H_∞ optimal control or sub-optimal control was developed by several authors^[1-10]. The discussion of these problems was almost focused on output-feedback or measurement feedback^[2-4] and state-feedback^[1,5,6]. Specially, [3] shows the concept of disturbance attenuation in the sense of L_2 norms and gives a set of sufficient condition for the problem of local disturbance attenuation with internal stability in a nonlinear affine system via measurement feedback, that is when the set of measured variables is just a function of the state of plant and of the disturbance input (also see [2]), and in terms of the solution of a pair of Hamilton-Jacobi inequality, which is the nonlinear version of the Riccati inequality considered in analysis of the H_∞ (sub) optimal control problem for linear

systems. [2] presents a necessary condition for the existence of a solution to the problem of (local) disturbance attenuation in the case of measurement feedback, and yields the construction of a feedback law for disturbance attenuation. The necessity of these sufficient conditions has also been discussed in [4] and [9]. [10] discusses the H_∞ -controller design problem in case of decentralized large scale nonlinear systems by using the Hamilton-Jacobi inequality approach via measurement feedback. The solution to the output feedback H_∞ control for nonlinear systems was obtained in [4] and [9]. The state feedback H_∞ control has also been investigated in [5] and [6].

More recent contributions to this area of research are the works [6] and [10]. In particular, [6] presents the problem of H_∞ control for nonlinear systems with a known Lyapunov function, and using dissipation inequality a state feedback controller is designed to grant that the closed-loop system is globally asymptotically stable. The most advantage of this method is not rely on solution of any HJIE's (Hamilton-Jacobi-Isaacs equation).

* Foundation item: supported by National Nature Science Foundation of China (59975033).

Received date: 2000-01-18; Revised date: 2001-09-13.

In this paper we addressed, along the research line of [6] and [10], nonlinear H_∞ -control problem for periodic disturbance attenuation (or, periodical nonlinear H_∞ -control problem PNHP). Using the notion of dissipativeness and Lyapunov function, some disturbance attenuation for the solution of H_∞ control problem are obtained and the suitable state controller is constructed explicitly.

2 Notation

Definition 2.1 If the signal $w(t)$ satisfying

$$w(t) = a_0/2 + \sum_{i=1}^N (a_i \cos i\omega t + b_i \sin i\omega t) = a_0/2 + \sum_{i=1}^N (a_i \cos i\omega t + b_i \sin i\omega t) + R(i\omega t) \approx a_0/2 + \sum_{i=1}^N (a_i \cos i\omega t + b_i \sin i\omega t), \quad (1)$$

where $\lim_{t \rightarrow \infty} R(i\omega t) = 0$, we call $w(t)$ is a periodical signal.

For all internal stable systems, the system gain for

$w(t) = \sum_{i=0}^l w_i(t)$ as input signal can be approximately expressed as follows^[10]

$$\gamma_{\max} = \max_{t \geq 0} \sqrt{\frac{\int_0^t [z_1^2(\tau) + \dots + z_l^2(\tau)] d\tau}{\int_0^t [w_0^2(\tau) + \dots + w_l^2(\tau)] d\tau}}, \quad (2)$$

where $z_i (i = 0, 1, \dots, l)$ are the output of disturbance.

Consider nonlinear system Σ

$$\Sigma: \dot{x} = f(x) + \sum_{i=0}^l g_{w_i}(x) w_i + g(x) u, \quad (3)$$

$$z = h(x) + d(x) u, \quad (4)$$

where $x \in \mathbb{R}^n$ stands for the state, $u \in \mathbb{R}^m$ is the local control input, $w_i \in \mathbb{R}^r (i = 0, \dots, l)$ as Definition 2.1, are the square-integrable exogenous disturbances, $z \in \mathbb{R}^q$ is an output to be regulated, $f(x), g(x)$ are smooth vector fields in \mathbb{R}^n and \mathbb{R}^m . $g_{w_i}(x)$ is a matrix for weight of exogenous disturbances, $h(x), d(x)$ are all known smooth mapping vector functions of appropriate dimensions, and satisfying $f(0) = 0, h(0) = 0$. For convenience, we denote $g_w(x) = [g_{w_0}, \dots, g_{w_l}]$, $w = [w_0^T, \dots, w_l^T]^T$, such that Σ can be written as

$$\Sigma: \dot{x} = f(x) + g_w(x) w + g(x) u, \quad (5)$$

$$z = h(x) + d(x) u. \quad (6)$$

Definition 2.2 Suppose for the initial state of sys-

tem Σ is $x_0 = 0$ and a neighborhood of $X \subseteq \mathbb{R}^n$ of x_0 , for any $\hat{x} \in X$; there exist a t_1 and a $w(t)$ such that $\hat{x} = \Phi(t_1, 0, x, w(t))$. If this happens, then called that the Σ is state reachable at x_0 . Where the function $\Phi(t_1, 0, x, w(t))$ is the solution of equation $\dot{x} = f(x) + g_w(x) w + g(x) u$ at $x_0 = 0$ and $u = 0$.

Definition 2.3 The pair $\{f, h\}$ is said to be locally detectable if there exists a neighborhood X of the point $x = 0$ such that, if $x(t)$ is any integral curve of $\dot{x} = f(x)$ satisfying $f(0) = 0, h(0) = 0$.

The PNHP to be addressed in this paper is as follows.

Given a scalar $\gamma \geq \gamma_{\max} > 0$, design a locally smooth nonlinear state feedback control game $u = \alpha(x)$ for Σ , with $\alpha(0) = 0$, and such that:

1) The origin is locally asymptotically stable equilibrium point of the closed-loop system;

2) the L_2 -gain of the closed-loop system from w to z is not larger than γ , i.e., satisfies the following inequality

$$\int_0^t z^T(\tau) z(\tau) d\tau \leq \gamma^2 \int_0^t w^T(\tau) w(\tau) d\tau, \text{ for all } t \geq 0.$$

3 Main result

Lemma 3.1 If system Σ at $x_0 = 0$ is state reachable, and there exists a differentiable smooth positive definite matrix function $V(x)$, satisfying $V(0) = 0$, so $(\partial V / \partial x) f(x) + h^T(x) h(x) + (\partial V / \partial x) g_w(x) g_w^T(x) (\partial V / \partial x)^T / (4\gamma^2) \leq 0$. (7)

Proof Since the L_2 -gain of the system is less than or equal to γ and the system is reachable, for any $T_1 \geq t_0$, we have

$$0 \leq \int_{t_0}^{T_1} (\gamma^2 w^T(\tau) w(\tau) - z^T(\tau) z(\tau)) d\tau < \infty. \quad (8)$$

That is

$$\int_{t_0}^{T_1} (\gamma^2 w^T(\tau) w(\tau) - z^T(\tau) z(\tau)) d\tau \geq - \int_0^{T_1} (\gamma^2 w^T(\tau) w(\tau) - z^T(\tau) z(\tau)) d\tau. \quad (9)$$

We order

$$V(x) = - \lim_{T_1 \rightarrow \infty} \inf_{w \in L_2[0, T_1]} \int_0^{T_1} (\gamma^2 w^T(\tau) w(\tau) - z^T(\tau) z(\tau)) d\tau, \quad (10)$$

consider the nonlinear system is reachable at $x_0 = 0$, when $w(t) \equiv 0$, the $x(t) = x_0$ is obviously the solution

of $z(t) \equiv 0$, so $V(x) \geq 0$.

Obviously

$$V(x) \leq \int_{t_0}^0 (\gamma^2 w^T(\tau) w(\tau) - z^T(\tau) z(\tau)) d\tau < \infty, \quad (11)$$

when $x_0 = 0$, $V(0) = 0$.

Consider the optimal problem for system Σ

$$W(x) = \inf_{w(\tau)} \int_0^\infty (\gamma^2 w^T(\tau) w(\tau) - z^T(\tau) z(\tau)) d\tau. \quad (12)$$

It can be known from optimal control theory that $W(x(t), t)$ satisfies

$$\begin{aligned} \frac{\partial W}{\partial x} = & \\ & - \min_{w(t)} \{ \gamma^2 w^T w - z^T z + \frac{\partial W}{\partial x} (f(x) + g_w(x) w) \}. \end{aligned} \quad (13)$$

It also can be said that $W(x(t), t)$ satisfies

$$\begin{aligned} \frac{\partial W}{\partial x} = & \\ & - \min_{w(t)} \{ \gamma^2 w^T w - z^T z + \frac{\partial W}{\partial x} (f(x) + g_w(x) w) \}. \end{aligned} \quad (14)$$

Because, $W(x) = -V(x)$, that is $\frac{\partial W}{\partial x} = -\frac{\partial V}{\partial x}$, and $\frac{\partial W}{\partial t} = 0$, so that

$$\begin{aligned} & - \min_{w(t)} \{ \gamma^2 w^T w - z^T z + \frac{\partial W}{\partial x} (f(x) + g_w(x) w) \} = \\ & - \min_{w(t)} \{ \gamma^2 w^T w - z^T z - (\frac{\partial V}{\partial x}) f(x) - (\frac{\partial V}{\partial x}) g_w(x) w \} = \\ & - \min_{w(t)} \{ \gamma^2 w^T w - h^T(x) h(x) - \frac{\partial V}{\partial x} f(x) - \\ & \frac{1}{4\gamma^2} \frac{\partial V}{\partial x} g_w(x) g_w^T(x) (\frac{\partial V}{\partial x})^T - \\ & \frac{1}{4\gamma^2} \frac{\partial V}{\partial x} g_w(x) g_w^T(x) (\frac{\partial V}{\partial x})^T - \frac{\partial V}{\partial x} g_w(x) w \} = \\ & \min_{w(t)} \{ \| \gamma w - \frac{1}{2\gamma} g_w^T(x) (\frac{\partial V}{\partial x})^T \|^2 + [\frac{\partial V}{\partial x} f(x) + \\ & h^T(x) h(x) + \frac{1}{4\gamma^2} \frac{\partial V}{\partial x} g_w(x) g_w^T(x) (\frac{\partial V}{\partial x})^T] \} = 0. \end{aligned}$$

We denote

$$\begin{aligned} P(x) = & \frac{\partial V}{\partial x} f(x) + h^T(x) h(x) + \\ & \frac{1}{4\gamma^2} \frac{\partial V}{\partial x} g_w(x) g_w^T(x) (\frac{\partial V}{\partial x})^T. \end{aligned} \quad (15)$$

Obviously

$$\| \gamma w - \frac{1}{2\gamma} g_w^T(x) (\frac{\partial V}{\partial x})^T \|^2 \geq 0,$$

so that

$$\begin{aligned} P(x) = & \frac{\partial V}{\partial x} f(x) + h^T(x) h(x) + \\ & \frac{1}{4\gamma^2} \frac{\partial V}{\partial x} g_w(x) g_w^T(x) (\frac{\partial V}{\partial x})^T \leq 0. \end{aligned}$$

Q.E.D.

Lemma 3.2 Suppose that the system Σ is state reachable at $x_0 = 0$, if there exist a neighborhood $X \subseteq \mathbb{R}^n$ and a differentiable smooth positive definite matrix function $V(x)$ for $x(0) \in X$, and suppose system Σ satisfies the following hypothesis.

Assumption 1 There exists a smooth function $V(x) \geq 0$ in the neighborhood of X such that $\frac{\partial V}{\partial x} \leq 0$.

Assumption 2 Matrix function $d(x)$ is of full-column rank for any $x \in \mathbb{R}^n$.

Assumption 3 Function $V(x)$ is a diagonal matrix. Then given any positive scalar $\gamma \geq \gamma_{\max} > 0$, the control law

$$u = -[d^T(x)d(x)]^{-1} [g^T(x)(\partial V/\partial x)^T/2 + d^T(x)h(x)] \quad (16)$$

solves the H_∞ control problem for this system.

Proof For convenience we denote $Q(x) = [d^T(x)d(x)]^{-1}$, evidently $Q(x) > 0$.

We know from Assumption 1, the dissipation inequality of system satisfies

$$\begin{aligned} & V(x(t_1)) - V(x(t_0)) + \\ & \int_{t_0}^{t_1} L(x(\tau), w(\tau), u(\tau)) d\tau \leq 0 \text{ for any } t_1 \geq t_0, \end{aligned}$$

where

$$L(x, w, u) = \| h(x) + d(x)u \|^2 - \gamma^2 \| w \|^2, \quad (17)$$

$$\begin{aligned} H(x, w, u) = & (\frac{\partial V}{\partial x}) \dot{x} + L(x, w, u) = \\ & \frac{\partial V}{\partial x} (f(x) + g_w(x)w + g(x)u) + \\ & (h(x) + d(x)u)^T (h(x) + d(x)u) - \gamma^2 w^T w = \\ & \frac{\partial V}{\partial x} f(x) + \frac{\partial V}{\partial x} g(x) + \frac{\partial V}{\partial x} g(x)u(x) + \\ & h^T(x)h(x) + u^T(x)d^T(x)h(x) + \\ & h^T(x)d(x)u(x) + \\ & u^T(x)d^T(x)d(x)u(x) - \gamma^2 w^T w. \end{aligned}$$

From Assumption 1, the system satisfies the condition of saddle point, such that for w and u from $\partial H/\partial w = 0$, $\partial H/\partial u = 0$ we can get the saddle point of $H(x, w, u)$.

$$w_p = \frac{1}{2\gamma^2} g_w^T(x) \left(\frac{\partial V}{\partial x} \right)^T, \quad (18)$$

$$u_p = - (d^T(x) d(x))^{-1} \cdot \frac{1}{2} g^T(x) \left(\frac{\partial V}{\partial x} \right)^T + d^T(x) h(x). \quad (19)$$

Consider (18) and (19), we obtain

$$\begin{aligned} H(x, w_p, u_p) &= \left(\frac{\partial V}{\partial x} \right) \dot{x} + L(x, w_p, u_p) = \\ &= \frac{\partial V}{\partial x} [f(x) + \frac{1}{2\gamma^2} g_w(x) g_w^T(x) \left(\frac{\partial V}{\partial x} \right)^T - \\ &\quad \frac{1}{2} [d^T(x) d(x)]^{-1} g(x) g^T(x) \left(\frac{\partial V}{\partial x} \right)^T - \\ &\quad [d^T(x) d(x)]^{-1} g(x) d^T(x) h(x)] + h^T(x) h(x) - \\ &\quad \frac{1}{4\gamma^2} \left(\frac{\partial V}{\partial x} \right) g_w(x) g_w^T(x) \left(\frac{\partial V}{\partial x} \right)^T - \\ &\quad \frac{1}{2} [d^T(x) d(x)]^{-1} [g^T(x) \left(\frac{\partial V}{\partial x} \right)^T + \\ &\quad d^T(x) h(x)]^T d^T(x) h(x) - \\ &\quad \frac{1}{2} [d^T(x) d(x)]^{-1} d^T(x) h(x) [g^T(x) \left(\frac{\partial V}{\partial x} \right)^T + \\ &\quad d^T(x) h(x)] + [d^T(x) d(x)]^{-1} \left[\frac{1}{2} g^T(x) \left(\frac{\partial V}{\partial x} \right)^T + \right. \\ &\quad \left. d^T(x) h(x) \right]^T \cdot \left[\frac{1}{2} g^T(x) \left(\frac{\partial V}{\partial x} \right)^T + d^T(x) h(x) \right] = \\ &= \frac{\partial V}{\partial x} f(x) + h^T(x) h(x) + \\ &\quad \frac{1}{4\gamma^2} \frac{\partial V}{\partial x} g_w(x) g_w^T(x) \left(\frac{\partial V}{\partial x} \right)^T - \\ &\quad \frac{1}{4} Q(x) \frac{\partial V}{\partial x} g(x) g^T(x) \left(\frac{\partial V}{\partial x} \right)^T - \\ &\quad Q(x) \frac{\partial V}{\partial x} g(x) d^T(x) h(x) - \\ &\quad Q(x) h^T(x) d(x) d^T(x) h(x) = \\ &= P(x) - Q(x) \left\| \frac{1}{2} \frac{\partial V}{\partial x} g(x) + d^T(x) h(x) \right\|^2. \end{aligned}$$

Since $Q(x) = [d^T(x) d(x)]^{-1} > 0$, so that

$$Q(x) \left\| \frac{1}{2} \frac{\partial V}{\partial x} g(x) + d^T(x) h(x) \right\|^2 \geq 0.$$

$$\begin{aligned} H(x, w_p, u_p) &\leq \frac{\partial V}{\partial x} f(x) + h^T(x) h(x) + \\ &\quad \frac{1}{4\gamma^2} \frac{\partial V}{\partial x} g_w(x) g_w^T(x) \left(\frac{\partial V}{\partial x} \right)^T. \end{aligned}$$

From Lemma 3.1 we can obtain

$$H(x, w_p, u_p) \leq 0.$$

Q.E.D.

4 Conclusion

This paper has considered the problem of nonlinear H_∞ -control problem for periodic disturbance attenuation. We have developed a methodology for designing nonlinear state feedback controllers that ensure local stability and a prescribed bound for the closed-loop system. The proposed design methodology involves the explicit construction of the control law on a Lyapunov function of system, which avoids the need for solving HJIE's. There is real value in the work of this paper.

References

- [1] Van der Schaft A J. A state-space approach to nonlinear H_∞ control [J]. Systems & Control Letters, 1992, 16(1): 1-8
- [2] Isidori A and Kang W. H_∞ control via measurement feedback for general nonlinear systems [J]. IEEE Trans. Automat. Contr., 1995, 40(3): 466-472
- [3] Isidori A and Kang W. Disturbance attenuation and H_∞ -control via measurement feedback in nonlinear systems [J]. IEEE Trans. Automat. Contr., 1992, 37(9): 1283-1293
- [4] Ball J A, Helton J W and Walker M L. H_∞ control for nonlinear systems with output feedback [J]. IEEE Trans. Automat. Contr., 1993, 38(4): 546-558
- [5] Shen T and Tamura K. Robust H_∞ -control of uncertain nonlinear systems via state feedback [J]. IEEE Trans. Automat. Contr., 1995, 40(4): 766-768
- [6] Su W, Souza C E de and Xie L. H_∞ control for asymptotically stable nonlinear systems [J]. IEEE Trans. Automat. Contr., 1999, 44(5): 989-993
- [7] Seo J H, Jo C H and Lee S H. Decentralized H_∞ -controller design [J]. Automatica, 1999, 35(5): 865-876
- [8] Tuan H D and Hosoe S. On linear robust H_∞ controllers for a class of nonlinear singular perturbed systems [J]. Automatica, 1999, 35(4): 753-759
- [9] Isidori A. A necessary condition for nonlinear H_∞ control via measurement feedback [J]. Systems & Control Letters, 1994, 23(3): 169-177
- [10] Yang G H, Wang J, Soh C B and Lam J. Decentralized H_∞ -controller design for nonlinear systems [J]. IEEE Trans. Automat. Contr., 1999, 44(3): 578-583

本文作者简介

李 曦 1965年生, 讲师, 华中理工大学博士生, 主要研究方向: 非线性控制理论及应用, 控制系统设计。

唐小琦 1958年生, 副教授, 博士, 主要研究方向: 非线性控制理论及应用, 计算机过程控制, 控制系统设计。

周云飞 1956年生, 教授, 博士生导师, 主要研究方向: 数控加工技术, 计算机过程控制。

陈吉红 1965年生, 教授, 博士, 主要研究方向: 数控加工技术, 计算机过程控制, 控制系统设计。